

Compressed sensing with prior information

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Compressed sensing: A revolution in sampling theory

- ▶ During the last 7 years, we have been witnessing a **revolution in sampling theory**.
- ▶ **Main conclusion:** sparse signals can be recovered from very few, “what appears to be incomplete” measurements in a tractable way.
- ▶ Initiated by the works of Donoho, and of Candès and Tao (~ 2004).
- ▶ Opened up a new field called **compressed sensing** or **compressive sampling**: Very active area. To follow:
 - Compressive sensing resources at <http://dsp.rice.edu/cs>
 - Nuit-Blanche Blog at <http://nuit-blanche.blogspot.com>
- ▶ Relies heavily on the theory of **sparse approximations** – around for more than two decades (transforms such as wavelets, curvelets, Gabor,...).
- ▶ Interesting and difficult mathematics and important applications such as seismic signal processing, imaging, and inversion.

Sampling and Reconstruction: Big Picture

Inherently analog signals: Audio, images, seismic, etc.

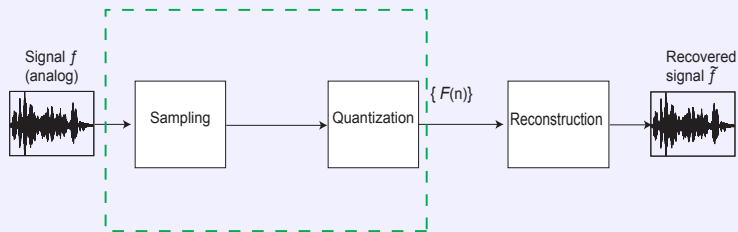
Objective: Use digital technology to store and process analog signals – find efficient digital representation of analog signals.

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(Generalized) Sampling of Analog Signals



SIGNAL ACQUISITION
Scheme based on **signal model**

SIGNAL RECOVERY
Based on both
acquisition method and
signal model

Outline of the talk

- ▶ Sampling theory and compressed sensing – an overview
- ▶ The use of prior information in compressed sensing
 - ▶ in the reconstruction stage: recovery via weighted ℓ_1 minimization
 - ▶ in the sampling stage: adaptive compressed sensing

Sampling and Reconstruction

Objectives:

- ▶ **Sampling scheme.** Specify how to obtain finitely many measurements of the signal f from which one can “recover” f . That is, the acquired samples should contain **sufficient information to recover/approximate f** .
- ▶ **Quantization.** Specify how to digitize the sample values (crucial for A/D conversion) in a way that is robustly implementable in analog hardware.
- ▶ **Reconstruction scheme.** Specify how to recover f from the samples.

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Main Problems and Challenges

- ▶ Must **model the signals of interest**, e.g., bandlimited, sparse etc... Note that without modeling, there is no hope of a “sampling theory”.
- ▶ Specify when we have **exact recovery**.
- ▶ When we don't have exact recovery, **tie the resolution of the approximation to the sampling density** (i.e., grid size, total number of samples etc.).
- ▶ Quantization has its own challenges, e.g., see work by Saab, OY et al.
- ▶ In any case, the schemes must be **practicable**.

Classical Sampling Theorem

We all know the “classical sampling theorem” of Shannon, Nyquist, Whittaker, Kotelnikov, Ogura, Borel, even Cauchy...

Signal model. The set of all bandlimited functions with bandlimit Ω – denote this set by B_Ω .

Sampling scheme. Collect values of $f \in B_\Omega$ on a **sufficiently dense uniform grid**, i.e., $\{f(n\tau) : n = \dots, -2, -1, 0, 1, 2, \dots\}$. Specifically, $\tau < \frac{1}{2\Omega}$.

Reconstruction scheme. Exact reconstruction via

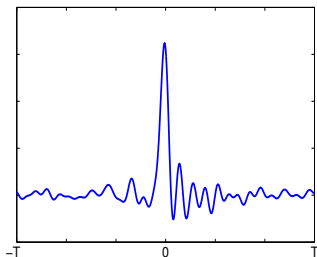
$$f(t) = \sum_n f(n\tau)\phi(t - n\tau), \quad \forall t.$$

Here ϕ is an appropriate low-pass filter.

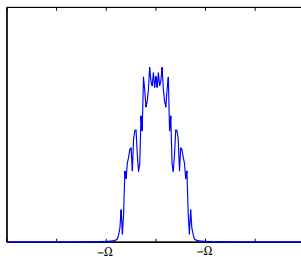
Practicability. If we slightly oversample, we can use a filter ϕ with fast decay, so obtain local reconstruction. This way, we also get robustness w.r.t. noise.

Classical Sampling Theorem: The picture

A bandlimited f

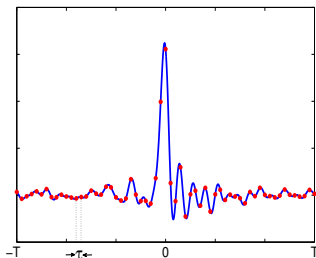


Fourier transform of f

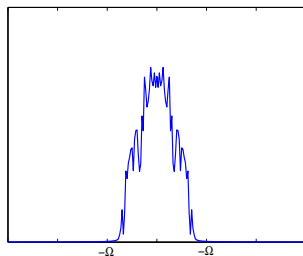


Classical Sampling Theorem: The picture

A bandlimited f



Fourier transform of f



Need $N \approx 2\Omega \times 2T$ samples to reconstruct f on $[-T, T]$.

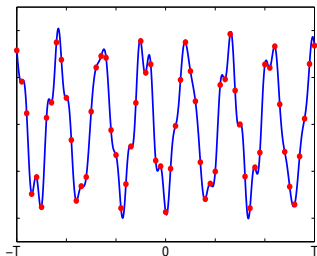
Equivalently: Every bandlimited function $f \in B_\Omega$, restricted to $[-T, T]$, can be represented by a vector $\mathbf{f} \in \mathbb{R}^N$ which we obtain by collecting N measurements.

Compressive Sampling Theory

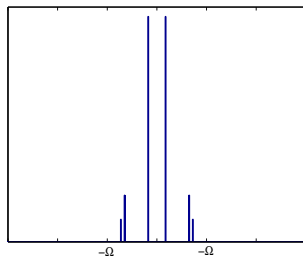
Above: Reduced a bandlimited function f to a vector \mathbf{f} in \mathbb{R}^N .

Question: Can we reduce the dimensionality of the problem by restricting the signal class further?

Another bandlimited f



Fourier transform of f



Do we still need $N \approx 4\Omega T$ samples to reconstruct $\mathbf{f} \in \mathbb{R}^N$?

Compressive Sampling Theory

Rephrase the question. Suppose we have:

Signal model. $f \in B_\Omega$ and f has a **sparse Fourier transform**.

Still need to sample at the Nyquist rate for a good (perfect) reconstruction?

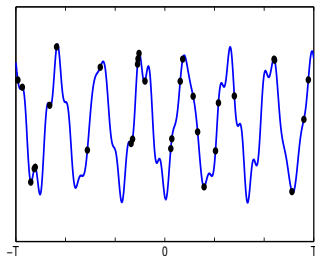
Compressive Sampling Theory

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New sampling scheme. Consider the following set of $m < N$ samples at (random) irregular points.



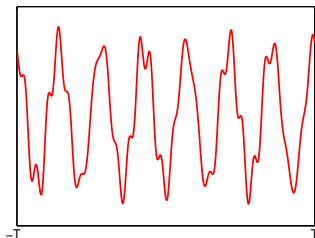
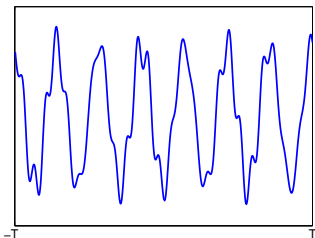
Average sampling density is **only 50% of Nyquist rate**, i.e., $m \approx N/2$.

Claim: We can recover f from these samples!

Recovery scheme. Find signal with matching samples that has the “sparsest” Fourier transform.

Compressive Sampling Theory

Here is the reconstruction obtained from the above samples (approx. 50% of Nyquist rate)



- ▶ We get essentially perfect reconstruction!
- ▶ How did we reconstruct? Next...

Compressive Sampling Theory – general framework

- ▶ Signal $f \in \mathbb{R}^N$. We want to collect information on f (*in the example: f is the full signal.*).
- ▶ **Model the signal class:** f admits a sparse representation w.r.t. a **known basis B** : $f = B^*x$ where x is **sparse**. (*in the example: B is the Fourier basis.*)
- ▶ **Specify a measurement scheme:** Construct an $m \times N$ **measurement matrix M** with $m \ll N$

$$\mathbf{f}_{\text{meas}} = Mf = MB^*x$$

(*in the example: \mathbf{f}_{meas} is the vector of non-uniform samples and M is the random restriction matrix in the example.*)

- ▶ **Reconstruction method:** Solve the underdetermined **sparse recovery problem**:

$$x_{\text{approx}} = \text{“sparsest” } z \text{ such that } \mathbf{f}_{\text{meas}} = MB^*z.$$

Compressive Sampling Theory: main questions

Sparse recovery problem:

$x_{\text{approx}} =$ “sparsest” z such that $\mathbf{f}_{\text{meas}} = MB^*z$.

Main questions:

1. How do we find the **sparsifying basis** B ?
2. How do we construct the **measurement matrix** M ?
3. **How many measurements** do we need to have $x_{\text{approx}} = x$?
4. How do we **solve** the sparse recovery problem?

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Compressive Sampling Theory - sparsity transforms

First address question 1: How do we find sparsity transforms?

- ▶ Note that this is dependent heavily on the **class of signals** of interest.
- ▶ In the above example, the sparsity transform was Fourier transform.
- ▶ **Applied and computational harmonic analysis** community has been developing such transforms during the last three decades that are tailored to important signal classes such as: audio, natural images, seismic data and images.
- ▶ Rich area with interesting mathematics, directly applicable constructive results such as **wavelet transform, curvelet transform** etc.
- ▶ Next, we give examples of some important sparsity transforms.

Sparsity transform - natural images

Wavelet transform sparsifies natural images.

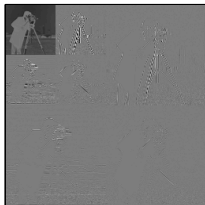
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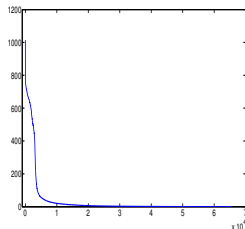
a wavelet atom



wavelet transform



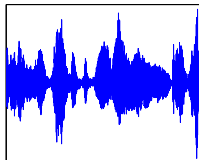
sorted coefficients



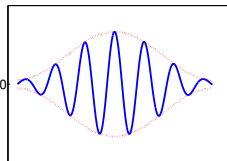
Sparsity transform - audio

Short-time Fourier (Gabor) transform sparsifies audio signals.

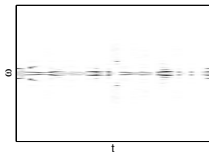
audio signal



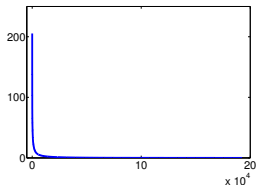
a Gabor atom



STFT transform



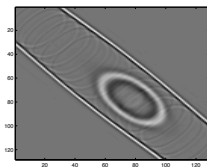
sorted coefficients



Sparsity transform - seismic

Curvelet transform sparsifies seismic data and images.

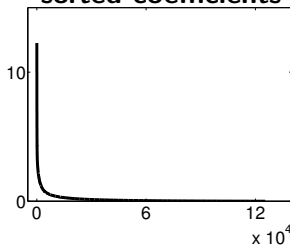
sampled Green's function



a curvelet atom



sorted coefficients



Compressive Sampling Theory: main questions

Sparse recovery problem:

x_{approx} = “sparsest” z such that $\mathbf{f}_{\text{meas}} = MB^*z$.

Main questions:

1. How do we find the sparsifying basis B ?
2. How do we construct the **measurement matrix** M ?
3. **How many measurements** do we need to have $x_{\text{approx}} = x$?
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Reconstruction: sparse recovery problem

Want to reconstruct f from the measurements

$$b = Mf = \underbrace{MB^*}_A x.$$

Some design goals:

1. **Exact recovery for all sufficiently sparse signals.** Want to recover **every** k -sparse x from the measurements $b = Ax$ with n as small as possible (say we fix k , N).
2. **Close to the best k -term approximation for compressible signals.** Want good estimates if x is not sparse but can be well-approximated by a sparse signal.
3. **Robustness to noise in either case above.** Want good estimates when the measurements are contaminated by noise, i.e., $\hat{b} = Ax + e$ where e is additive noise.
4. **Computationally tractable recovery method.**

CS – many surprises since 2004!

We can achieve all the goals above (main results by Donoho, and Candes, Romberg, Tao) – just use a recovery algorithm based on ℓ_1 minimization:

$$\Delta_1(b) := \arg \min \|z\|_1 \quad \text{subject to } Az = b \quad \text{no noise case}$$

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In particular:

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There are other algorithms for CS recovery—e.g., Δ_p with $0 < p < 1$, OMP, CoSamp, ...

How to choose the measurement matrix

- ▶ There are **precise conditions** on A (in terms of its RIP constants) that guarantee that the above results hold.
- ▶ For example, if A is a **random matrix with iid Gaussian entries**, then

$$m \gtrsim k \log(N/k)$$

will suffice. **Num. of measurements scales only logarithmically with the ambient dimension: grid size in our previous example.**

- ▶ This is **theoretically optimal** (deep results in geometric functional analysis).
- ▶ Other classes (Bernoulli, partial Fourier, ...) of random matrices will do, too!

Choosing the measurement matrix — more remarks

- ▶ Gaussian and sub-Gaussian matrices are **unitarily invariant**, so the dimension relation is independent of the sparsity basis. These are **universal measurement matrices**:

M is Gaussian and B is unitary $\implies A = MB^*$ is Gaussian.

- ▶ Ideal for **dimension reduction in simulations**. Also, acquisition with **simultaneous sources**.

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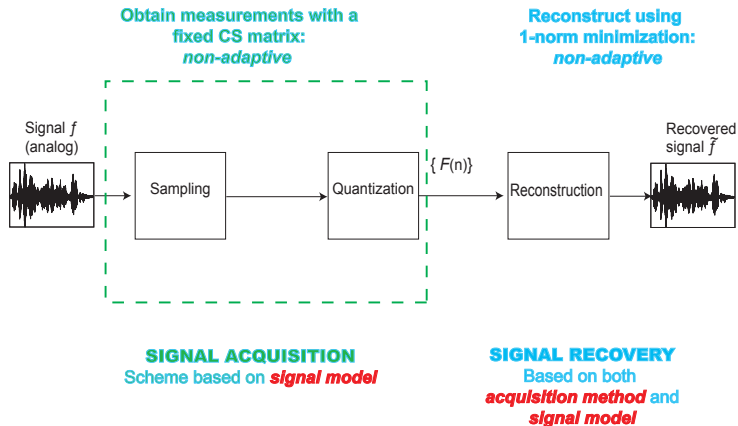
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- ▶ Ideal for **dimension reduction in simulations**. Also, acquisition with **simultaneous sources**.
- ▶ Difficult to implement depending on the physics—e.g., in the sampling example. In such cases:
 - ▶ sample in a domain that is **incoherent** with the sparsity domain: e.g.,
 $\text{sparse in Fourier} \implies \text{sample in time}$
 - ▶ Randomly sub-sample (possibly on a jittered grid), i.e., “apply” a restriction matrix R .

The corresponding $A = RM$ will be a “good” compressive sampling matrix.

CS – incorporating prior info

Note that, like classical sampling, CS is a non-adaptive sampling paradigm:



CS – incorporating prior info

Remainder of the talk: We investigate methods of incorporating prior information on the support of the specific signal of interest to our sampling and reconstructions schemes.

1. Recovery using weighted ℓ_1 minimization.

- ▶ **Sensing is non-adaptive:** Collect the measurements b (or \hat{b} if there is noise) using an arbitrary CS matrix. On the other hand:
- ▶ **Recovery is adaptive:** Use prior support info to come up with better recovery methods.

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2. Adaptive Compressed Sensing.

- ▶ **Sampling scheme is adaptive** and incorporates a “**compressive antialiasing**” stage.
- ▶ **Recovery is also adaptive** (using weighted ℓ_1).

Part I: CS – recovery via weighted ℓ_1

Setting: Suppose we have prior information on the support of x . In particular we have a **support estimate** that is generally **partial** and possibly **inaccurate**.

Want: **incorporate such information in the recovery algorithm** to get better results than those obtained via ℓ_1 minimization.

Why is this relevant?

Signals with Prior Information

- ▶ In many applications, it is possible to draw an estimate of the support of the signal, for example:

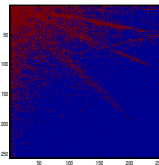
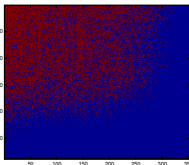
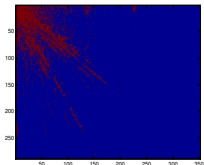
Signals with Prior Information

- ▶ In many applications, it is possible to draw an estimate of the support of the signal, for example:
 - ▶ Natural images have large DCT coefficients that are localized in the low frequency subbands.



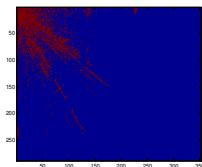
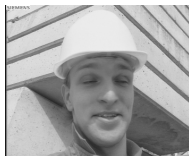
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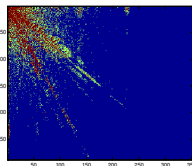
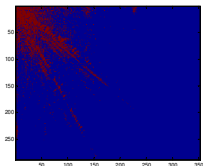
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- ▶ In many applications, it is possible to draw an estimate of the support of the signal, for example:
 - ▶ Natural images have large DCT coefficients that are localized in the low frequency subbands.
 - ▶ Video sequences are temporally correlated, resulting in a shared subset of their support.



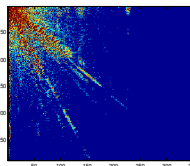
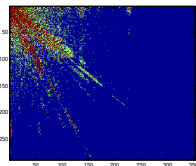
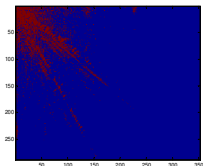
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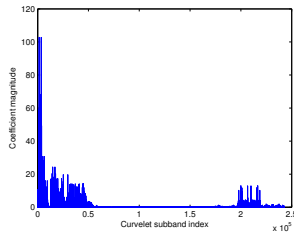
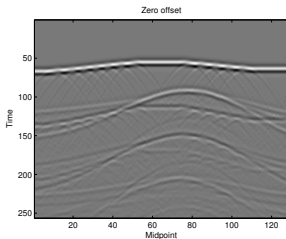
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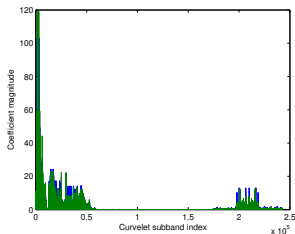
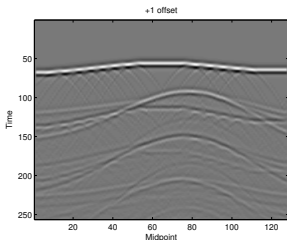
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 - ▶ Natural images have large DCT coefficients that are localized in the low frequency subbands.
 - ▶ Video sequences are temporally correlated, resulting in a shared subset of their support.
 - ▶ Other signals such as seismic data, ...



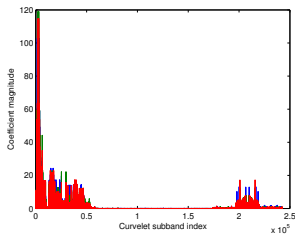
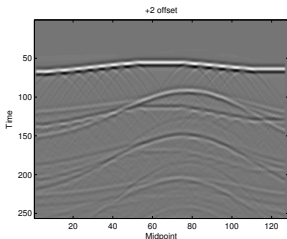
Signals with Prior Information

- ▶ In many applications, it is possible to draw an estimate of the support of the signal, for example:
 - ▶ Natural images have large DCT coefficients that are localized in the low frequency subbands.
 - ▶ Video sequences are temporally correlated, resulting in a shared subset of their support.
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Problem formulation

Suppose that x is a k -sparse signal with unknown support T_0 .

Given:

1. CS measurements of x (i.e., $b = Ax$, or $\hat{b} = Ax + e$ with $\|e\|_2 \leq \epsilon$).
2. A partially accurate support estimate \tilde{T} . Let's quantify—two important parameters:

$$\rho := \frac{\#\tilde{T}}{\#T_0} \quad \text{relative size of the estimated support}$$

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In general, we have $0 \leq \rho \leq \frac{N}{k}$ and $0 \leq \alpha \leq \min\{1, \frac{1}{\rho}\}$.

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Goals:

- ▶ Incorporate \tilde{T} into the recovery algorithm (to get better recovery),
- ▶ Obtain theoretical recovery guarantees depending on the size and accuracy of \tilde{T} (i.e., ρ and α).

Proposed Algorithm – weighted ℓ_1 minimization

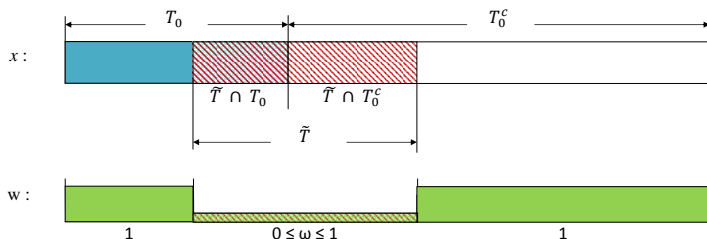
Given a set of (noisy) measurements \hat{b} , define

$$\Delta_{1,w}^\epsilon(\hat{b}) := \arg \min_x \|x\|_{1,w} \text{ subject to } \|Ax - \hat{b}\|_2 \leq \epsilon$$

where

$$w_i = \begin{cases} 1, & i \in \tilde{T}^c, \\ \omega, & i \in \tilde{T}, \end{cases} \text{ for some } 0 \leq \omega \leq 1.$$

Above $\|x\|_{1,w} := \sum_i w_i |x_i|$, and $\|e\|_2^2 \leq \epsilon$.



Improved sufficient conditions for weighted ℓ_1

We prove the following theorem in the case of weighted ℓ_1 :

Theorem [FMSY]

Suppose for some $a > \max\{1, (1 - \alpha)\rho\}$, $\delta_{ak} + a\gamma\delta_{(a+1)k} < a\gamma - 1$. Then

$$\|\Delta_{1,w}^\epsilon(\hat{b}) - x\|_2 \leq C'_0\epsilon + C'_1k^{-1/2}(\omega\|x_{T_\epsilon^c}\|_1 + (1 - \omega)\|x_{\tilde{T}^c \cap T_\epsilon^c}\|_1)$$

where $\gamma = \left(\omega + (1 - \omega)\sqrt{1 + \rho - 2\alpha\rho}\right)^{-2}$.

Remarks.

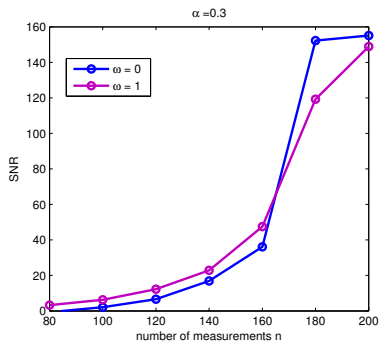
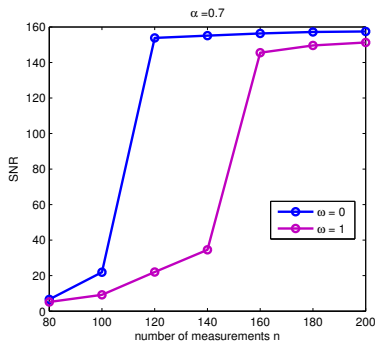
1. Above, $0 \leq \omega \leq 1$ is a fixed weight. If we set $\omega = 1$, our theorem reduces to the robust recovery theorem of CRT.
2. Recall $0 \leq \alpha \leq 1$ describes the accuracy of \tilde{T} and ρ describes its size.
3. **The sufficient conditions above are weaker than those for ℓ_1 minimization iff $\alpha > 0.5$. (Same holds for the constants.)**
4. Earlier work on the case $\omega = 0$: e.g., Borries, Vaswani and Lu; Jacques. Our results, to our knowledge, provide weakest sufficient cond. and smallest constants.

Numerical Experiments – sparse signals

- ▶ SNR averaged over 20 experiments for k -sparse signals x with $k = 40$, and $N = 500$.

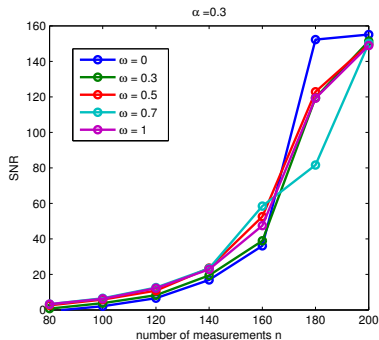
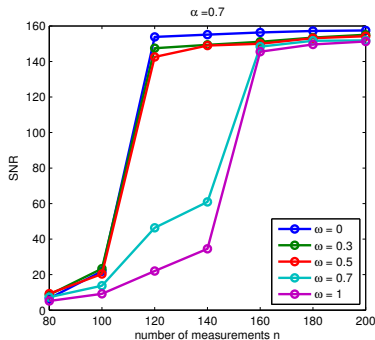
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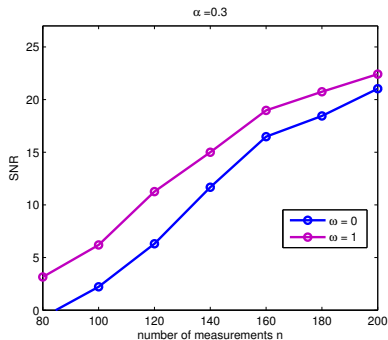
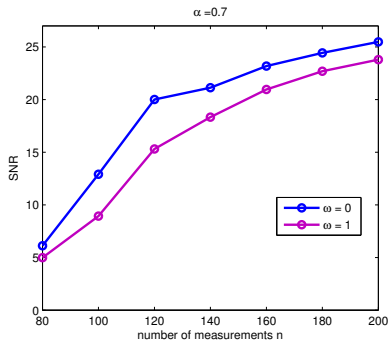
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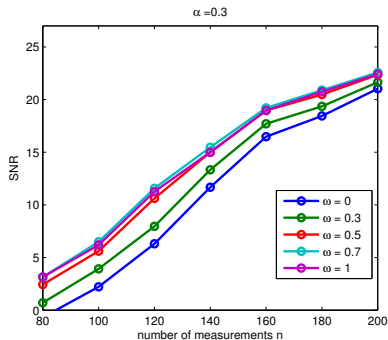
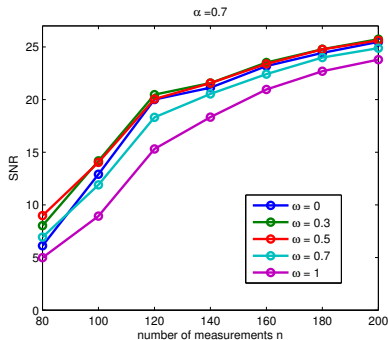
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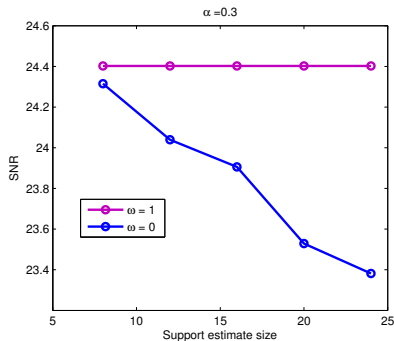
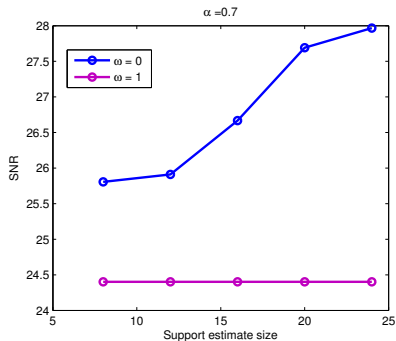


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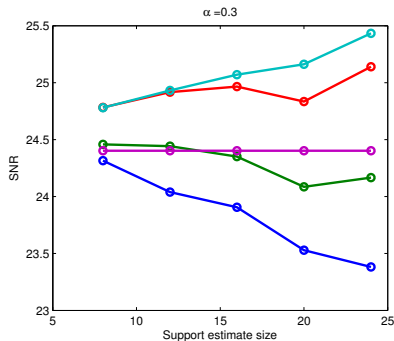
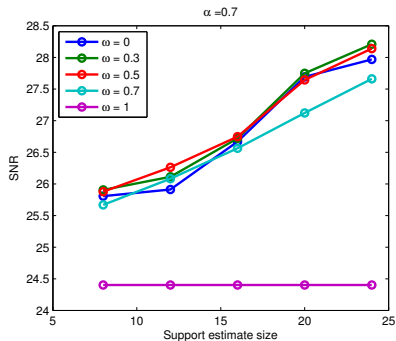
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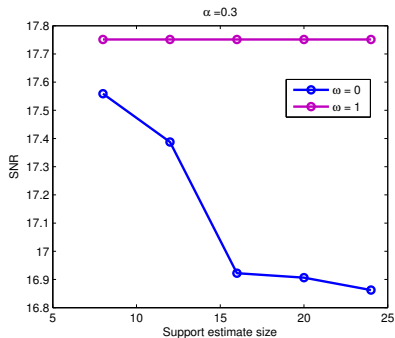
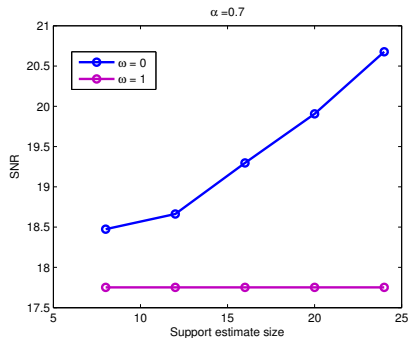
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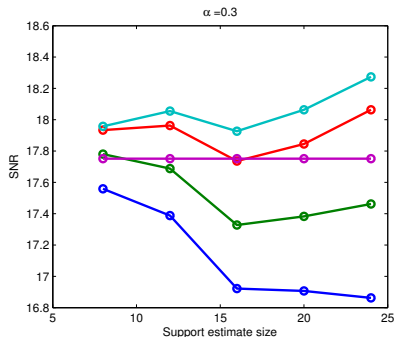
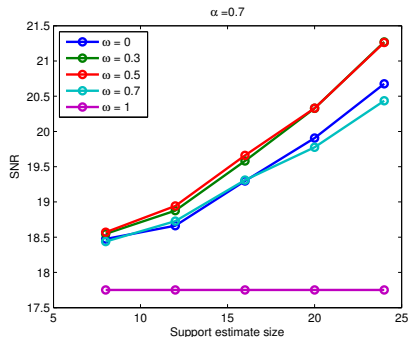
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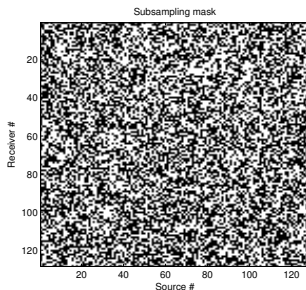
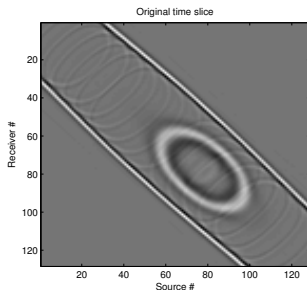
- ▶ As ω goes to zero,
 - ▶ the constant $C'_1(\omega)$ increases
 - ▶ the term $\omega \|x_{T_0^c}\|_1 + (1 - \omega) \|x_{\tilde{T}^c \cap T_0^c}\|_1$ decreases
- ▶ There may exist $0 < \omega < 1$ that minimizes their product (depending on the signal class as well as properties of the measurement matrix A).

Application: Compressed sensing of seismic lines

- ▶ Full seismic line with 128 shots, 128 receivers, and 256 time samples.

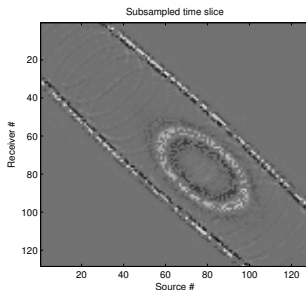
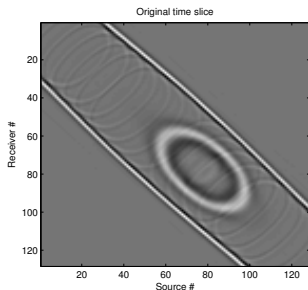
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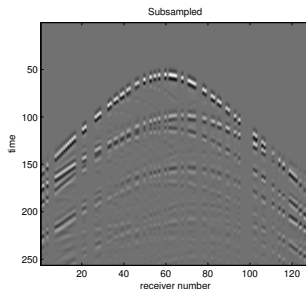
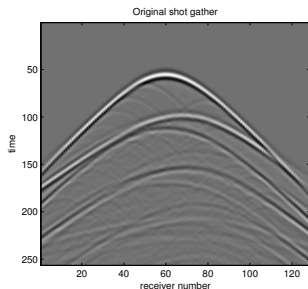
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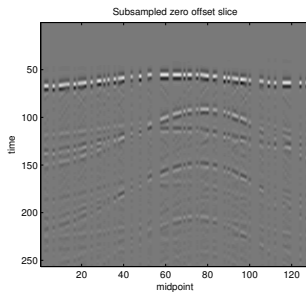
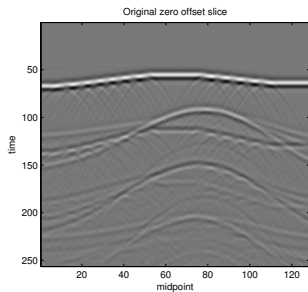
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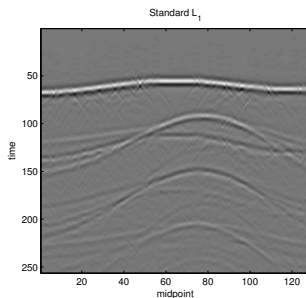
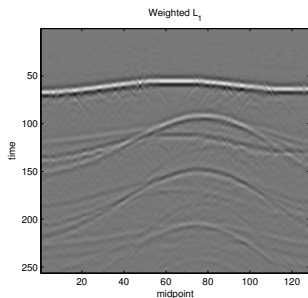
Recovery in offset domain

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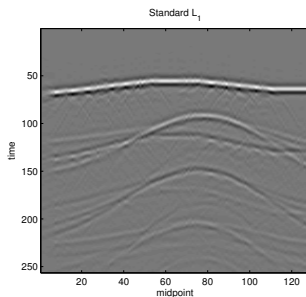
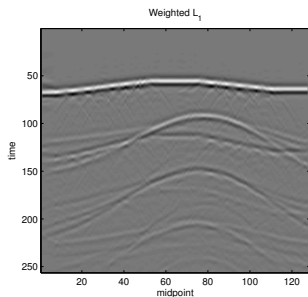
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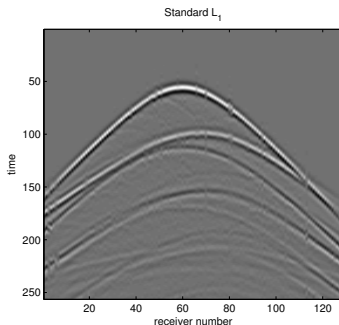
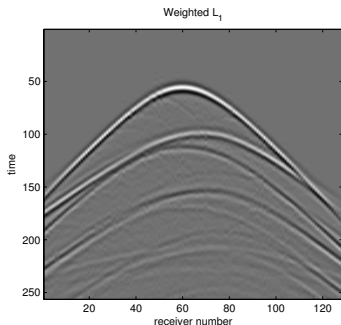
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- ▶ Use the support of the zero offset slice to weight the recovery of other offset slices (eg: +2 offset).



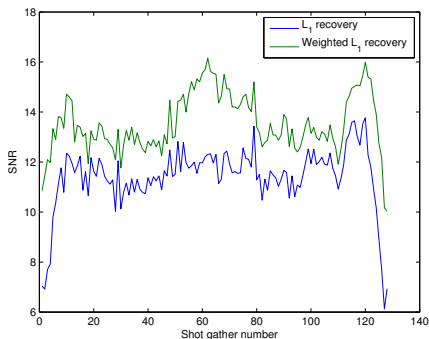
Performance of weighted ℓ_1 vs standard ℓ_1

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- ▶ Map the data back to the source receiver domain (eg: shot gather 60).
- ▶ Signal to noise ratio (SNR) of all 128 shot gathers.



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 - ▶ How would an iterative weighted ℓ_1 algorithm with **fixed** weights perform compared to IRL1 of Boyd and Candès?

Part II: Adaptive CS

Disclaimer. This is very recent work and the results we will present are preliminary.

Main Problem. “Compressive aliasing” ... To explain, recall the classical sampling theory setup:

Ideal signal model: Bandlimited with known bandwidth, say maximum frequency ω_1 .

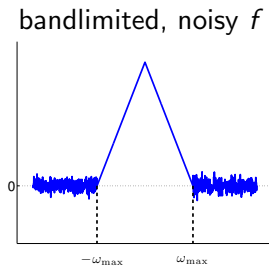
Sampling scheme: Sample with frequency $\omega_s \geq 2\omega_1$

In practice: Signals may have higher bandwidth $\omega_{\max} > \frac{\omega_s}{2}$; moreover there may be off-band noise.

Problem: This would cause aliasing as well as noise in the sampled & reconstructed signal.

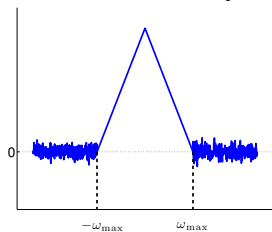
Remedy: Use a front-end low-pass filter and **force the signal to obey the ideal signal model**. Resulting approximation is the best approximation with bandwidth $\omega_s/2$.

Antialiasing in classical sampling

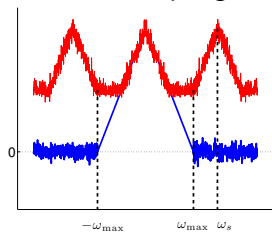


Antialiasing in classical sampling

bandlimited, noisy f

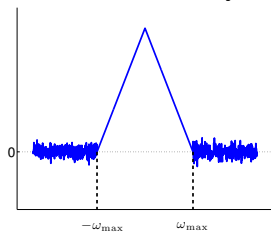


after sampling

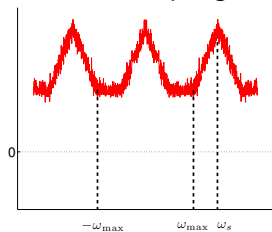


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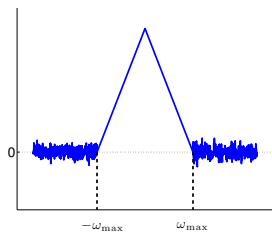


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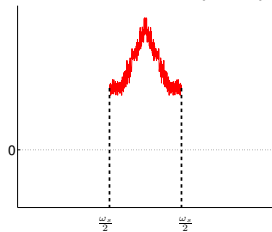


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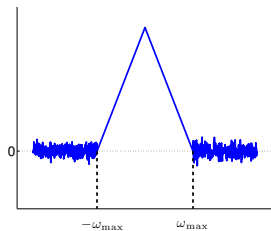


reconstruction (LPF)

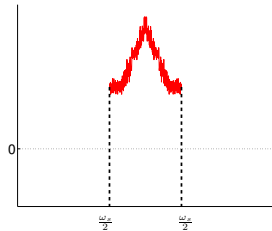


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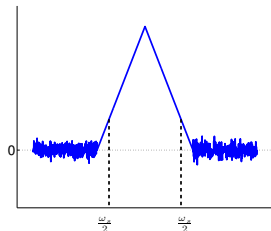
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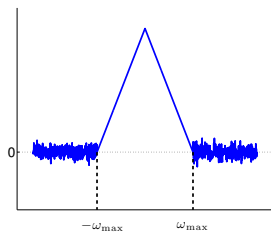


front-end LPF

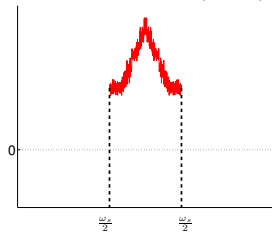


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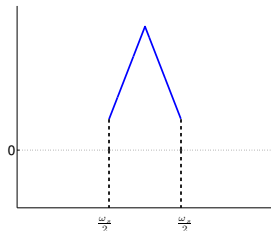
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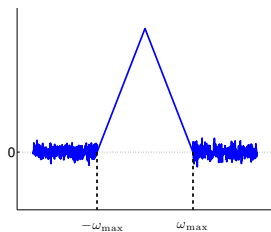


front-end LPF

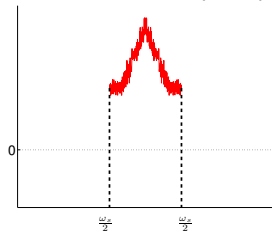


Antialiasing in classical sampling

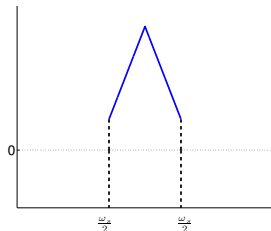
bandlimited, noisy f



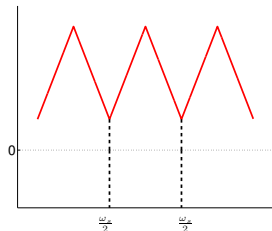
reconstruction (LPF)



front-end LPF

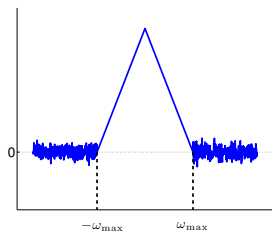


after sampling

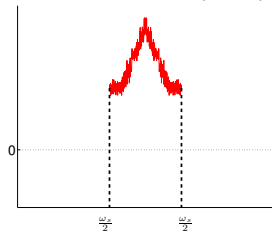


Antialiasing in classical sampling

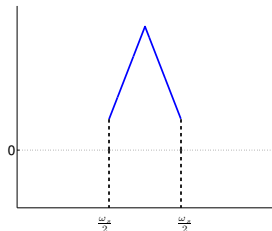
bandlimited, noisy f



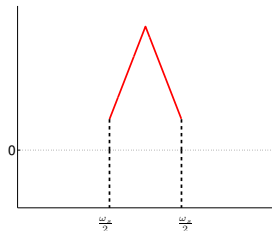
reconstruction (LPF)



front-end LPF



reconstruction



Compressive aliasing

Parallel to our discussion above:

- ▶ **Ideal signal model in CS:** k -sparse signals in \mathbb{R}^N ($k \sim$ bandwidth)
- ▶ **Sampling scheme:** Hit the signal by an appropriate $m \times N$ measurement matrix ($m \sim$ sampling density, is suited to recover all k -sparse signals)

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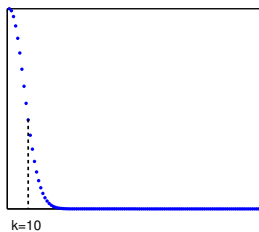
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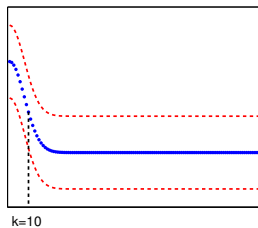
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- ▶ **Good news:** CS robust recovery theorems guarantee that the approximation is “almost” as good as the best we could hope for.
- ▶ **Problem:** If the coefficient decay is not fast enough, large coefficients can still get significantly distorted : **compressive aliasing**.
- ▶ **Goal:** Find an antialiasing method when there is prior support information (as in Part I).

Compressive aliasing

compressible x

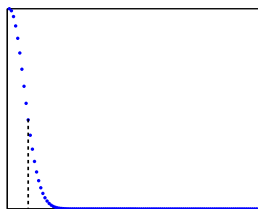


after CS and recovery



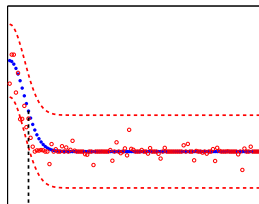
Compressive aliasing

compressible x



$k=10$

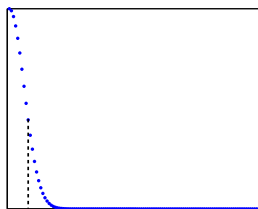
after CS and recovery



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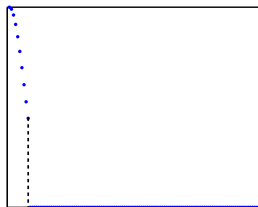
Compressive aliasing

compressible x



$k=10$

“sparse-filtered” x



$k=10$

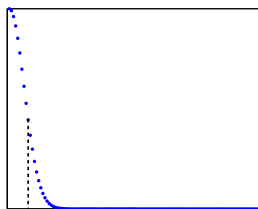
after CS and recovery



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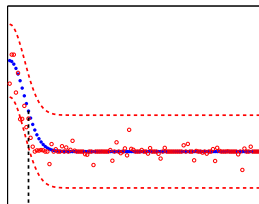
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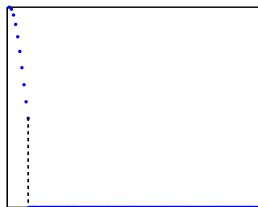
$k=10$

after CS and recovery

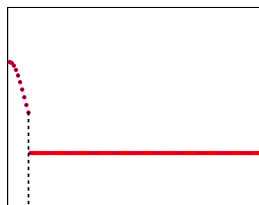


$k=10$

“sparse-filtered” x



$k=10$



$k=10$

Moral: If we knew the support of the sparse signal, we could sparsify it (analogous to low-pass filtering in classical sampling) and force it to obey the signal model ideal for CS.

Problem: If we knew the support, we don't need CS!

Compromise: What if we have a partial and possibly inaccurate support estimate?

Adaptive CS setup

Let f be a signal that is compressible with respect to a transform B , i.e.,

$$f = B^* x, \quad x \text{ decays fast.}$$

Suppose we knew $T_k = \text{supp}(x_k)$ – indices of largest k coefficients of x .

“sparse-filter” f : $f_{\text{sp}} = B^* W^2 \underbrace{Bf}_x$.

Here W is a diagonal matrix such that $W_{jj} = \begin{cases} 1 & \text{if } j \in T_k \\ 0 & \text{if } j \notin T_k \end{cases}$

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The corresponding CS scheme—with CS matrix A —would be:

$$A \underbrace{B^* W^2 Bf}_{f_{\text{sp}}} = y \quad \leftarrow \text{measurements of "sparse-filtered" } f$$

and we can recover x_k (thus $f_k = B^* x_k$) by solving

$$\min \|z\|_1 \text{ subject to } AB^* z = y.$$

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Adaptive CS – more realistic approach

Same setup as in the previous slide: $f = B^*x$, x compressible
However, instead of T_k , we have a **partial and inaccurate estimate** \tilde{T}_k .

Proposed Method.

- ▶ Let W be a diagonal weighting matrix such that

$$W_{jj} = \begin{cases} 1 & \text{if } j \in \tilde{T}_k \\ \omega & \text{if } j \notin \tilde{T}_k \end{cases}$$

where $0 < \omega < 1$ (some intermediate value for robustness).

- ▶ **Adaptive sampling:** Collect the (noisy) samples

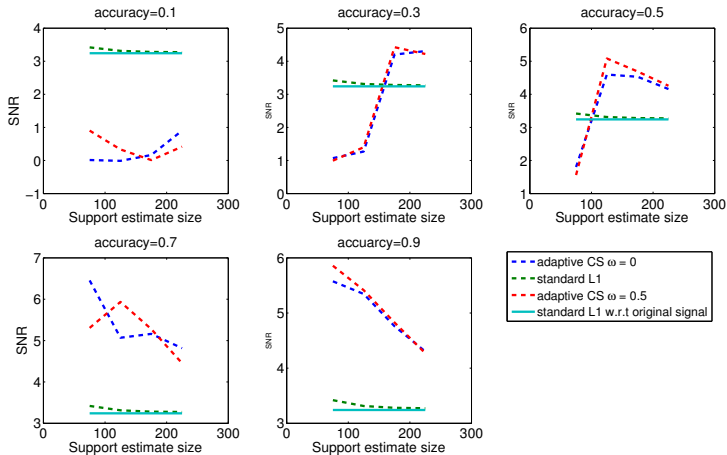
$$AB^*W^2Bf + e = y.$$

- ▶ **(Adaptive) Reconstruction** via

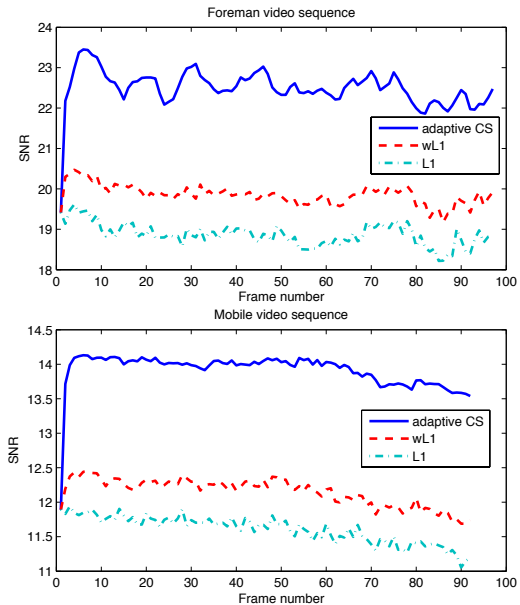
- ▶ ℓ_1 minimization: $\min \|z\|_1$ subject to $\|AB^*z - y\| \leq \epsilon$, or
- ▶ weighted ℓ_1 : $\min \|z\|_1$ subject to $\|AB^*Wz - y\| \leq \epsilon$.

Example with synthetic signals

Signals with coefficients decaying like $j^{-0.7}$; $N = 500$, $m = 50$.

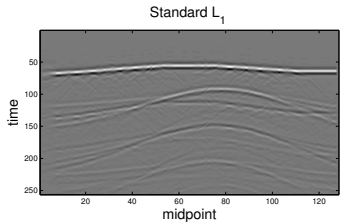
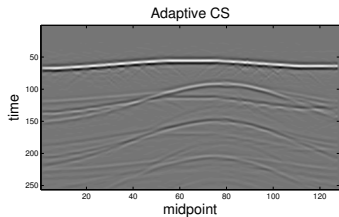
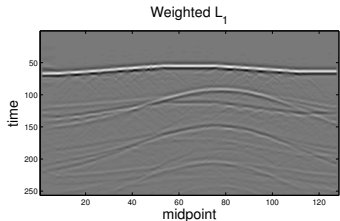
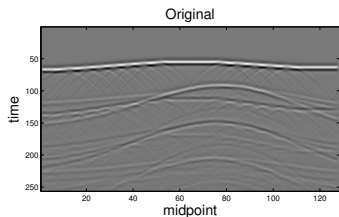


Example with video frame sequences



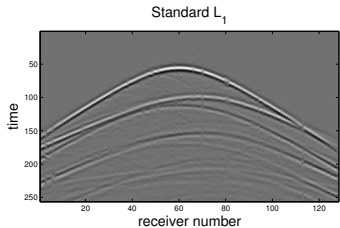
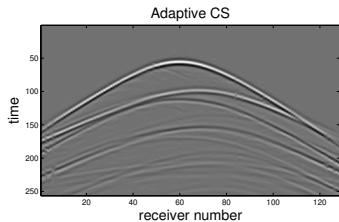
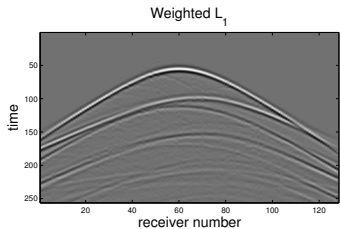
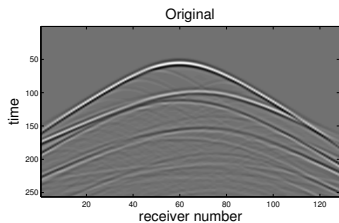
Example: adaptive CS of seismic lines

Same experiment as before – %50 of source/receiver pairs are missing.
Below, we show offset=+2...



Example: adaptive CS of seismic lines

Same experiment as before – %50 of source/receiver pairs are missing.
Below, we show shot gather 60...



Concluding remarks

- ▶ CS provides a powerful sampling theory for the acquisition of signals that admit a sparse or compressible representation in some transform domain.
- ▶ CS is a non-adaptive sampling paradigm.
- ▶ If prior information is available, it can be effectively used in both the sampling stage and the reconstruction stage.
- ▶ We are currently working on fine-tuning the adaptive CS approach and using it both for seismic acquisition and in computations.

Acknowledgements

This work was in part financially supported by the Natural Sciences and Engineering Research Council of Canada Discovery Grant (22R81254) and the Collaborative Research and Development Grant DNOISE II (375142-08). This research was carried out as part of the SINBAD II project with support from the following organizations: BG Group, BGP, BP, Chevron, ConocoPhillips, Petrobras, PGS, Total SA, and WesternGeco.